System Identification

Control Engineering EN, 3rd year B.Sc. Technical University of Cluj-Napoca Romania

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Method development

Matlab example

Accuracy guarantee

Multiple inputs and outputs 000000000

Part V

ARX identification

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Method development

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We stay in the single-output, single-input case for all but the last section.

Recall Types of models from Part I:

- Mental or verbal models
- Graphs and tables (nonparametric)
- Mathematical models, with two subtypes:
 - First-principles, analytical models
 - Models from system identification

The ARX method produces *parametric*, polynomial models.

Why ARX?			
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- General-order, fully implementable method with guarantees like correlation analysis
- Unlike correlation analysis, gives a *compact* model with a number of parameters proportional to the order of the system

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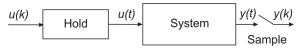
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Recall: Discrete time

We remain in the discrete-time setting:



In the ARX model structure, the output y(k) at the current discrete time step is computed based on previous input and output values:

$$y(k) + a_1y(k-1) + a_2y(k-2) + \dots + a_{na}y(k-na)$$

= $b_1u(k-1) + b_2u(k-2) + \dots + b_{nb}u(k-nb) + e(k)$
equivalent to
 $y(k) = -a_1y(k-1) - a_2y(k-2) - \dots - a_{na}y(k-na)$
 $b_1u(k-1) + b_2u(k-2) + \dots + b_{nb}u(k-nb) + e(k)$

e(k) is the noise at step k.

Model parameters: $a_1, a_2, \ldots, a_{n_a}$ and b_1, b_2, \ldots, b_{nb} .

Name: AutoRegressive (y(k) a function of previous y values) with eXogenous input (dependence on u)

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Polynomial rep	resentation		

Backward shift operator q^{-1} :

$$q^{-1}z(k)=z(k-1)$$

where z(k) is any discrete-time signal.

Then:

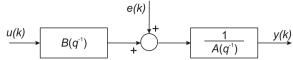
$$y(k) + a_1 y(k-1) + a_2 y(k-2) + \dots + a_{na} y(k-na)$$

= $(1 + a_1 q^{-1} + a_2 q^{-2} + \dots + a_{na} q^{-na}) y(k) =: A(q^{-1}) y(k)$
and:
 $b_1 u(k-1) + b_2 u(k-2) + \dots + b_{nb} u(k-nb)$
= $(b_1 q^{-1} + b_2 q^{-2} + \dots + b_{nb} q^{-nb}) u(k) =: B(q^{-1}) u(k)$

Therefore, the ARX model is written compactly:

$$A(q^{-1})y(k) = B(q^{-1})u(k) + e(k)$$

Or in graphical representation:



which holds because:

$$y(k) = \frac{1}{A(q^{-1})}[B(q^{-1})u(k) + e(k)]$$

Remark: The ARX model is quite general, it can describe arbitrary linear relationships between inputs and outputs. However, the noise enters the model in a restricted way, and later we introduce models that generalize this.

Linear regres			
Method development	Matlab example	Accuracy guarantee	Multiple inputs and outputs

Returning to the explicit representation:

$$y(k) = -a_1 y(k-1) - a_2 y(k-2) - \dots - a_{na} y(k-na)$$

$$b_1 u(k-1) + b_2 u(k-2) + \dots + b_{nb} u(k-nb) + e(k)$$

$$= [-y(k-1), \dots, -y(k-na), u(k-1), \dots, u(k-nb)]$$

$$\cdot [a_1, \dots, a_{na}, b_1, \dots, b_{nb}]^\top + e(k)$$

$$= :\varphi^\top(k)\theta + e(k)$$

So in fact ARX obeys the standard model structure in linear regression!

Vectors of regr	essors and pa	arameters	
Method development	Matlab example	Accuracy guarantee	Multiple inputs and outputs

Regressor vector: $\varphi(k) \in \mathbb{R}^{na+nb}$, previous output and input values. Parameter vector: $\theta \in \mathbb{R}^{na+nb}$, polynomial coefficients.

$$\varphi(k) = \begin{bmatrix} -y(k-1) \\ \vdots \\ -y(k-na) \\ u(k-1) \\ \vdots \\ u(k-nb) \end{bmatrix} \qquad \theta = \begin{bmatrix} a_1 \\ \vdots \\ a_{na} \\ b_1 \\ \vdots \\ b_{nb} \end{bmatrix}$$

A remark on vector notation

All vector variables are column by default, following standard control notation. Often we write them as transposed rows, to save vertical space:

$$Q = egin{bmatrix} q_1 \ q_2 \ q_3 \end{bmatrix} = [q_1, q_2, q_3]^ op$$

Note also: $Q^{\top} = [q_1, q_2, q_3].$

For instance, in the ARX formula:

$$y(k) = [-y(k-1), \dots, -y(k-na), u(k-1), \dots, u(k-nb)]$$
$$\cdot [a_1, \dots, a_{na}, b_1, \dots, b_{nb}]^\top + e(k) = \varphi^\top(k)\theta + e(k)$$

 $\varphi(k)$ is originally column, but must be transformed into a row to make the inner product work. So we transpose it: $\varphi^{\top}(k)$, obtaining the row [-y(k-1),...] (note here there is no transpose). On the other hand, θ must be a column in the inner product, hence it is not transposed; however, for convenience we write it as a transposed row, i.e. $[a_1,...]^{\top}$ (note the transpose!).

Consider now that we are given a dataset u(k), y(k), k = 1, ..., N, and we have to find the model parameters θ .

Then for any k:

$$\mathbf{y}(\mathbf{k}) = \varphi^{\top}(\mathbf{k})\theta + \varepsilon(\mathbf{k})$$

where $\varepsilon(k)$ is now interpreted as an equation error (hence the changed notation).

Objective: minimize the mean squared error:

$$V(\theta) = \frac{1}{N} \sum_{k=1}^{N} \varepsilon(k)^2$$

Remark: When $k \le na$, nb, zero- and negative-time values for u and y are needed to construct φ . They can be taken equal to 0 (assuming the system is in zero initial conditions).

Method development	Matlab example	Accuracy guarantee	Multiple inputs and outputs
Linear system	of equations		
-		$u(0) \cdots u(1-nb)]$ $u(1) \cdots u(2-nb)]$	

$$y(N) = \begin{bmatrix} -y(N-1) & \cdots & -y(N-na) & u(N-1) & \cdots & u(N-nb) \end{bmatrix} \theta$$

$\begin{bmatrix} y(1) \\ y(2) \\ \vdots \\ y(N) \end{bmatrix} = \begin{bmatrix} -y(0) & \cdots & -y(1 - na) & u(0) & \cdots & u(1 - nb) \\ -y(1) & \cdots & -y(2 - na) & u(1) & \cdots & u(2 - nb) \\ \vdots & \vdots & \vdots & \vdots & \\ -y(N - 1) & \cdots & -y(N - na) & u(N - 1) & \cdots & u(N - nb) \end{bmatrix} \cdot \theta$

 $Y = \Phi \theta$

with notations $Y \in \mathbb{R}^N$ and $\Phi \in \mathbb{R}^{N \times (na+nb)}$.

Method development	Matlab example 000000	Accuracy guarantee	Multiple inputs and outputs
ARX solution			

From linear regression, to minimize $\frac{1}{2} \sum_{k=1}^{N} \varepsilon(k)^2$ the parameters are:

$$\widehat{\theta} = (\Phi^{\top} \Phi)^{-1} \Phi^{\top} Y$$

Since the new $V(\theta) = \frac{1}{N} \sum_{k=1}^{N} \varepsilon(k)^2$ is proportional to the criterion above, the same solution also minimizes $V(\theta)$.

Method development	Matlab example	Accuracy guarantee	Multiple inputs and outputs
Solution for la	ge dataset	S	

When the number of data points N is very large, the form above is impractical.

Better form in that case:

$$\Phi^{\top} \Phi = \sum_{k=1}^{N} \varphi(k) \varphi^{\top}(k), \quad \Phi^{\top} Y = \sum_{k=1}^{N} \varphi(k) y(k)$$
$$\Rightarrow \widehat{\theta} = \left[\sum_{k=1}^{N} \varphi(k) \varphi^{\top}(k) \right]^{-1} \left[\sum_{k=1}^{N} \varphi(k) y(k) \right]$$

(Recall the similar "trick" from linear regression.)

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 Solution for large datasets (continued)

Remaining issue: the sum of *N* terms can grow very large, leading to numerical problems: (matrix of very large numbers)⁻¹ · vector of very large numbers.

Solution: Normalize element values by diving them by *N*. In equations, *N* simplifies so it has no effect on the analytical development, but in practice it keeps the numbers reasonable.

$$\widehat{\theta} = \left[\frac{1}{N}\sum_{k=1}^{N}\varphi(k)\varphi^{\top}(k)\right]^{-1}\left[\frac{1}{N}\sum_{k=1}^{N}\varphi(k)y(k)\right]$$

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Using the ARX model

One-step ahead prediction: The true output sequence is known, so all the delayed signals are available and we can simply plug them in the formula, together with the coefficients taken from θ :

$$\hat{y}(k) = -a_1y(k-1) - a_2y(k-2) - \dots - a_{na}y(k-na)$$

 $b_1u(k-1) + b_2u(k-2) + \dots + b_{nb}u(k-nb)$

Signals at negative and zero time can be taken equal to 0.

Example: On day k - 1, predict weather for day k.

Simulation: True outputs unknown, so we must use previously simulated outputs. Each y(k - i) is replaced by $\hat{y}(k - i)$:

$$\hat{y}(k) = -a_1\hat{y}(k-1) - a_2\hat{y}(k-2) - \dots - a_{na}\hat{y}(k-na)$$

$$b_1u(k-1) + b_2u(k-2) + \dots + b_{nb}u(k-nb)$$

(predicted outputs at negative and zero time can also be taken 0.)

Example: Simulation of an aircraft's response to emergency pilot inputs, that may be dangerous to apply to the real system.

Special case of ARX: FIR

Setting
$$A = 1$$
 ($na = 0$) in ARX, we get:

$$egin{aligned} y(k) &= B(q^{-1})u(k) + e(k) = \sum_{j=1}^{nb} b_j u(k-j) + e(k) \ &= \sum_{j=0}^{M-1} h(j)u(k-j) + e(k) \end{aligned}$$

the FIR model from correlation analysis!

To see this, take nb = M - 1, and $b_j = h(j)$. Note h(0), the impulse response at time 0, is assumed 0 - i.e. system does not respond instantaneously to changes in input.

ARX:
$$A(q^{-1})y(k) = B(q^{-1})u(k) + e(k)$$

FIR: $y(k) = B(q^{-1})u(k) + e(k)$

Since ARX includes relationships between current and previous outputs, it will be sufficient to take orders *na* and *nb* equal to the order of the dynamical system.

FIR needs a sufficiently large order *nb* (or length *M*) to model the entire transient regime of the impulse response (in principle, we only recover the correct model as $M \rightarrow \infty$).

 \Rightarrow more parameters \Rightarrow more data needed to identify them.

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Method development

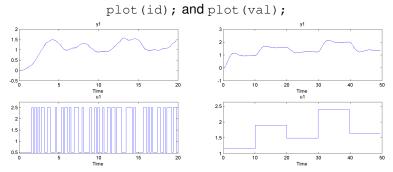
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Consider we are given the following, separate, identification and validation data sets.



Remarks: Identification input: a so-called *pseudo-random binary signal*. Validation input: a sequence of steps.

model = arx(id, [na, nb, nk]);

Arguments:

Identification data.

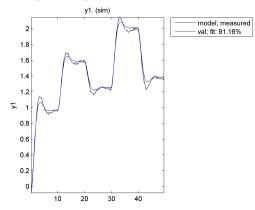
Array containing the orders of A and B and the delay nk.

Structure different from theory: includes explicitly a minimum delay *nk* between inputs and outputs, useful for systems with time delays.

$$\begin{aligned} y(k) + a_1 y(k-1) + a_2 y(k-2) + \ldots + a_{na} y(k-na) \\ &= b_1 u(k-nk) + b_2 u(k-nk-1) + \ldots + b_{nb} u(k-nk-nb+1) + e(k) \\ A(q^{-1}) y(k) &= B(q^{-1}) u(k-nk) + e(k), \text{ where:} \\ A(q^{-1}) &= (1+a_1q^{-1}+a_2q^{-2}+\ldots+a_{na}q^{-na}) \\ B(q^{-1}) &= (b_1+b_2q^{-1}+b_{nb}q^{-nb+1}) \end{aligned}$$

The theoretical structure is obtained by setting nk = 1. For nk > 1, we can also transform the new structure into the theoretical one by using a *B* polynomial of order nk + nb - 1, with nk - 1 leading zeros: $B_{\text{theor}}(q^{-1}) = 0q^{-1} + \dots 0q^{-nk+1} + b_1q^{-nk} + \dots + b_{nb}q^{-nk-nb+1}$ Method development Mattab example OCONCOLOGICO Multiple inputs and outputs OCONCOCOCO

Assuming the system is second-order, *in the ARX form*, and without time delay, we take na = 2, nb = 2, nk = 1. Validation: compare (model, val);



Results are quite bad.

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Structure sel	ection		

Alternate idea: try many different structures and choose the best one.

```
Na = 1:15;
Nb = 1:15;
Nk = 1:5;
NN = struc(Na, Nb, Nk);
V = arxstruc(id, val, NN);
```

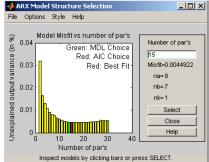
- struc generates all combinations of orders in Na, Nb, Nk.
- arxstruc identifies for each combination an ARX model (on the data in 1st argument), simulates it (on the data in the 2nd argument), and returns all the MSEs on the first row of V (see help arxstruc for the format of V).

To choose the structure with the smallest MSE:

```
N = selstruc(V, 0);
```

For our data, N = [8, 7, 1].

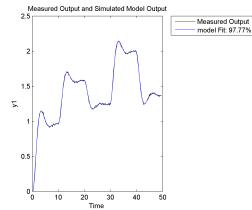
Alternatively, graphical selection: N = selstruc(V, 'plot'); Then click on bar corresponding to best (red) model and "Select", "Close".



(Later we learn other structure selection criteria than smallest MSE.)



model = arx(id, N); compare(model, val);



A better fit is obtained. However, 8th order systems are rare in real life, so something else is likely going on... we will see later.

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Main result			

Assumptions

• There exists a true parameter vector θ_0 so that:

$$y(k) = \varphi^{\top}(k)\theta_0 + v(k)$$

with v(k) a stationary stochastic process independent from u(k). **2** $E \{\varphi(k)\varphi^{\top}(k)\}$ is a nonsingular matrix.

Theorem

ARX identification is consistent: the estimated parameters $\hat{\theta}$ converge to the true parameters θ_0 , in the limit as $N \to \infty$.



Assumption 1 is equivalent to the existence of true polynomials A₀(q⁻¹), B₀(q⁻¹) so that:

$$A_0(q^{-1})y(k) = B_0(q^{-1})u(k) + v(k)$$

To motivate Assumption 2, recall

$$\widehat{\theta} = \left[\frac{1}{N}\sum_{k=1}^{N}\varphi(k)\varphi^{\top}(k)\right]^{-1}\left[\frac{1}{N}\sum_{k=1}^{N}\varphi(k)y(k)\right]$$

As $N \to \infty$, $\frac{1}{N} \sum_{k=1}^{N} \varphi(k) \varphi^{\top}(k) \to \mathrm{E} \left\{ \varphi(k) \varphi^{\top}(k) \right\}.$

- E {\(\varphi(k)\)\)\)\) is nonsingular if the data is "sufficiently informative" (e.g., u(k) should not be a simple feedback from y(k); see Söderström & Stoica for more discussion).
- E {φ(k)v(k)} = 0 e.g. if v(k) is white noise. Later on, we will discuss in more detail Assumption 3 and the role of E {φ(k)v(k)} = 0.

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MIMO syste	m		

So far we considered $y(k) \in \mathbb{R}$, $u(k) \in \mathbb{R}$, Single-Input, Single-Output (SISO) systems

Many systems are Multiple-Input, Multiple-Output (MIMO). E.g., aircraft. Inputs: throttle, aileron, elevator, rudder. Outputs: airspeed, roll, pitch, yaw.



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MIMO ARX			

Consider next y(k), $e(k) \in \mathbb{R}^{ny}$, $u(k) \in \mathbb{R}^{nu}$. MIMO ARX model:

$$A(q^{-1})y(k) = B(q^{-1})u(k) + e(k)$$
$$A(q^{-1}) = I + A_1q^{-1} + \ldots + A_{na}q^{-na}$$
$$B(q^{-1}) = B_1q^{-1} + \ldots + B_{nb}q^{-nb}$$

where *I* is the $ny \times ny$ identity matrix, $A_1, \ldots, A_{na} \in \mathbb{R}^{ny \times ny}$, $B_1, \ldots, B_{nb} \in \mathbb{R}^{ny \times nu}$.

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Concrete exa	ample		

Take na = 1, nb = 2, ny = 2, nu = 3. Then:

$$\begin{aligned} A(q^{-1})y(k) &= B(q^{-1})u(k) + e(k) \\ A(q^{-1}) &= I + A_1 q^{-1} \\ &= I + \begin{bmatrix} a_{1}^{11} & a_{1}^{12} \\ a_{1}^{21} & a_{1}^{22} \end{bmatrix} q^{-1} \\ B(q^{-1}) &= B_1 q^{-1} + B_2 q^{-2} \\ &= \begin{bmatrix} b_{1}^{11} & b_{1}^{12} & b_{1}^{13} \\ b_{1}^{21} & b_{1}^{22} & b_{1}^{23} \end{bmatrix} q^{-1} + \begin{bmatrix} b_{2}^{11} & b_{2}^{12} & b_{2}^{13} \\ b_{2}^{21} & b_{2}^{22} & b_{2}^{23} \end{bmatrix} q^{-2} \end{aligned}$$

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Concrete example (continued)

$$\begin{pmatrix} \begin{bmatrix} 1 & 0 \\ 0 & 1 \end{bmatrix} + \begin{bmatrix} a_1^{11} & a_1^{12} \\ a_1^{21} & a_1^{22} \end{bmatrix} q^{-1} \end{pmatrix} \begin{bmatrix} y_1(k) \\ y_2(k) \end{bmatrix}$$

$$= \begin{pmatrix} \begin{bmatrix} b_1^{11} & b_1^{12} & b_1^{13} \\ b_1^{21} & b_1^{22} & b_1^{23} \end{bmatrix} q^{-1} + \begin{bmatrix} b_2^{11} & b_2^{12} & b_2^{13} \\ b_2^{21} & b_2^{22} & b_2^{23} \end{bmatrix} q^{-2} \end{pmatrix} \begin{bmatrix} u_1(k) \\ u_2(k) \\ u_3(k) \end{bmatrix} + \begin{bmatrix} e_1(k) \\ e_2(k) \end{bmatrix}$$

Explicit relationship:

$$y_{1}(k) + a_{1}^{11}y_{1}(k-1) + a_{1}^{12}y_{2}(k-1)$$

$$= b_{1}^{11}u_{1}(k-1) + b_{1}^{12}u_{2}(k-1) + b_{1}^{13}u_{3}(k-1)$$

$$+ b_{2}^{11}u_{1}(k-2) + b_{2}^{12}u_{2}(k-2) + b_{2}^{13}u_{3}(k-2) + e_{1}(k)$$

$$y_{2}(k) + a_{1}^{21}y_{1}(k-1) + a_{1}^{22}y_{2}(k-1)$$

$$= b_{1}^{21}u_{1}(k-1) + b_{1}^{22}u_{2}(k-1) + b_{1}^{23}u_{3}(k-1)$$

$$+ b_{2}^{21}u_{1}(k-2) + b_{2}^{22}u_{2}(k-2) + b_{2}^{23}u_{3}(k-2) + e_{2}(k)$$

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Matlab exampl	е		

Consider a continuous stirred-tank reactor:

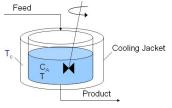


Image credit: mathworks.com

Input: coolant flow Q

Outputs:

- Concentration C_A of substance A in the mix
- Temperature *T* of the mix

Method development

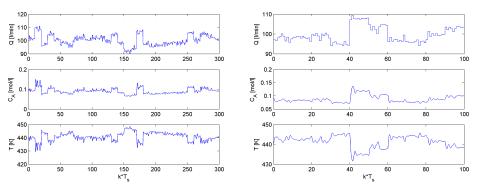
Matlab example

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Matlab: Experimental data

Left: identification, Right: validation



$$\begin{aligned} \mathcal{A}(q^{-1})y(k) &= \mathcal{B}(q^{-1})u(k) + \mathcal{B}(k) \\ \mathcal{A}(q^{-1}) &= \begin{bmatrix} a^{11}(q^{-1}) & a^{12}(q^{-1}) & \dots & a^{1ny}(q^{-1}) \\ a^{21}(q^{-1}) & a^{22}(q^{-1}) & \dots & a^{2ny}(q^{-1}) \\ \dots & a^{ny1}(q^{-1}) & a^{ny2}(q^{-1}) & \dots & a^{nyny}(q^{-1}) \end{bmatrix} \\ a^{ij}(q^{-1}) &= \begin{cases} 1 & \text{if } i = j \\ 0 & \text{otherwise} \end{cases} + a^{ij}_{1}q^{-1} + \dots + a^{ij}_{na_{ij}}q^{-na_{ij}} \\ B &= \begin{bmatrix} b^{11}(q^{-1}) & b^{12}(q^{-1}) & \dots & b^{1nu}(q^{-1}) \\ b^{21}(q^{-1}) & b^{22}(q^{-1}) & \dots & b^{2nu}(q^{-1}) \\ \dots & b^{ny1}(q^{-1}) & b^{ny2}(q^{-1}) & \dots & b^{nynu}(q^{-1}) \end{bmatrix} \\ b^{ij}(q^{-1}) &= b^{ij}_{1}q^{-nk_{ij}} + \dots + b^{ij}_{nb_{ij}}q^{-nk_{ij}-nb_{ij}+1} \end{aligned}$$

Arguments:

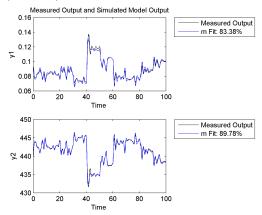
- Identification data.
- Matrices with orders of polynomials in A, B, and delays nk:

$$Na = \begin{bmatrix} na_{11} & \dots & na_{1ny} \\ \dots & & \\ na_{ny1} & \dots & na_{nyny} \end{bmatrix}$$
$$Nb = \begin{bmatrix} nb_{11} & \dots & nb_{1nu} \\ \dots & & \\ nb_{ny1} & \dots & nb_{nynu} \end{bmatrix}$$
$$Nk = \begin{bmatrix} nk_{11} & \dots & nk_{1nu} \\ \dots & & \\ nk_{ny1} & \dots & nk_{nynu} \end{bmatrix}$$

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Take na = 2, nb = 2, and nk = 1 everywhere in matrix elements:

Na = [2 2; 2 2]; Nb = [2; 2]; Nk = [1; 1]; m = arx(id, [Na Nb Nk]); compare(m, val);



Appendix: Nonlinear ARX (for project)

Nonlinear ARX structure

Recall standard ARX:

$$y(k) = -a_1 y(k-1) - a_2 y(k-2) - \dots - a_{na} y(k-na)$$

$$b_1 u(k-1) + b_2 u(k-2) + \dots + b_{nb} u(k-nb) + e(k)$$

Linear dependence on delayed outputs $y(k-1), \ldots, y(k-na)$ and inputs $u(k-1), \ldots, u(k-nb)$.

Nonlinear ARX (NARX) generalizes this to any nonlinear dependence:

$$y(k) = g(y(k-1), y(k-2), \dots, y(k-na), u(k-1), u(k-2), \dots, u(k-nb); \theta) + e(k)$$

Function *g* is parameterized by $\theta \in \mathbb{R}^n$, and these parameters can be tuned to fit identification data and thereby model a particular system.

Polynomial NARX

In our particular case, g is a polynomial of degree m in the delayed outputs and inputs:

$$y(k) = p(y(k-1), \dots, y(k-na), u(k-1), \dots, u(k-nb)) + e(k)$$

=: $p(d(k)) + e(k)$

where $d(k) = [y(k-1), \dots, y(k-na), u(k-1), \dots, u(k-nb)]^{\top}$ is the vector of delayed signals.

E.g., for orders na = nb = 1 (so $d(k) = [y(k-1), u(k-1)]^{\top}$) and degree m = 1, the model is:

$$y(k) = ay(k-1) + bu(k-1) + c + e(k)$$

which by further taking c = 0 recovers the linear ARX form

Polynomial NARX (continued)

For the same na = nb = 1 and degree m = 2:

$$y(k) = ay(k-1) + bu(k-1) + cy(k-1)^{2}$$
$$+ du(k-1)^{2} + wu(k-1)y(k-1) + z + e(k)$$

- Do not confuse with polynomial form $A(q^{-1})y(k) = B(q^{-1})u(k) + e(k)$
- The parameters are now the coefficients of the polynomial, e.g. $\theta = [a, b, c, d, w, z]^{\top}$
- Linear regression works as usual, finding the parameters that minimize the MSE!
- Negative and zero-time *y* and *u* can be taken 0, assuming system in zero initial conditions

Recall prediction versus simulation

One-step ahead prediction: True output sequence is known, delays vector d(k) is fully available:

$$\begin{aligned} x(k) = & [y(k-1), \dots, y(k-na), u(k-1), \dots, u(k-nb)]^\top \\ \hat{y}(k) = & g(d(k); \hat{\theta}) \end{aligned}$$

Simulation: True outputs unknown, use the previously simulated outputs to construct an *approximation* of d(k):

$$\widehat{x}(k) = [\widehat{y}(k-1), \dots, \widehat{y}(k-na), u(k-1), \dots, u(k-nb)]^{\perp}$$
$$\widehat{y}(k) = g(\widehat{d}(k); \widehat{\theta})$$